1.5MHz,30V,1.2AFPWMSynchronous Step-Down Converter

Features

- Wide Input Range: 4.5V to 30V
- Feedback Voltage: 0.8V±2%
- 1.5MHz Switch Frequency
- Up to 1.2A Output Current
- Force Constant Current PWM Operation
- COT control to achieve fast transient responses
- Integrated internal compensation
- Stable with Low ESR Ceramic Output Capacitors
- Over Current Protection with Hiccup Mode
- Thermal Shutdown
- Inrush Current Limit and Soft Start
- Build in Input Over Voltage Protection
- Available in SOT23-6 Package

Description

The SL2359 is a high efficiency 1.5MHz,synchronous step-down DC-DC converter capable of delivering up to 1.2A current with force PWM operation and adopts COT control with fast transient response performance. SL2359 integrates both high and low side switch with low $R_{DS(ON)}$ to minimize the conduction loss. Low output voltage ripple and small external inductor and frequency. capacitor size are achieved with1.5MHz switching The SL2359 requires a minimum number of readily available standard external components and is available in a 6-pin SOT23-6 RoHS compliant package.
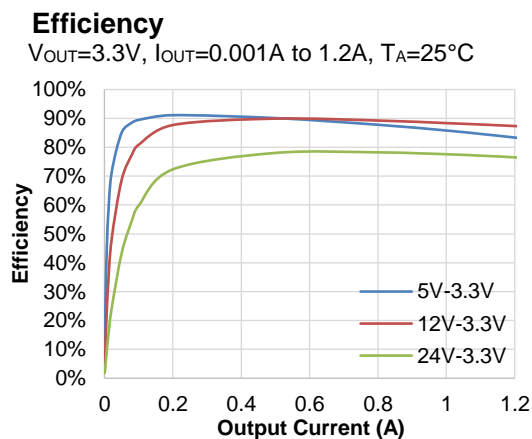
Application

- Security Equipment
- Distributed Power Systems
- Flat Panel Television and Monitors
- Smart Home
- Industrial Power Systems

Typical Application
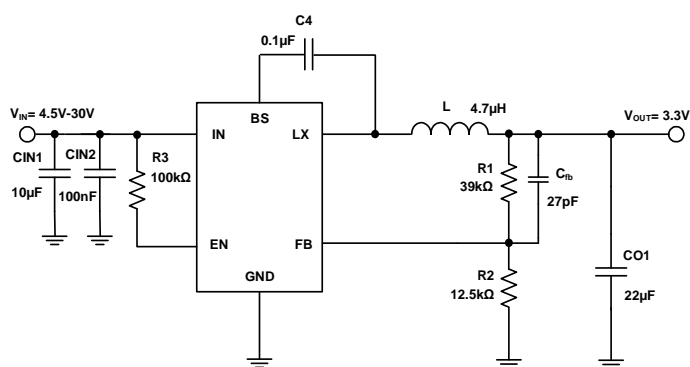

SL2359 **Typical Application Circuits (V_REF=0.8V)**

Efficiency
V_{OUT}=3.3V, I_{OUT}=0.001A to 1.2A, T_A=25°C

Absolute Maximum Ratings (Note 1)

Parameter	Min	Max	Unit
Input Supply Voltage	-0.3	33	V
Enable pin, EN	-0.3	V_{IN}	V
LX Voltages	-0.3	33	V
LX Voltages (<10ns transient)	-3.0	35	V
FB Voltage	-0.3	6	V
BS to LX Voltage	-0.3	6	V
Storage Temperature Range	-65	150	°C
Junction Temperature (Note2)		150	°C
Power Dissipation		1000	mW
Lead Temperature (Soldering, 10s)		260	°C

Package

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BS  [1]   ○      [6]  LX

GND [2]          [5]  IN

FB  [3]          [4]  EN
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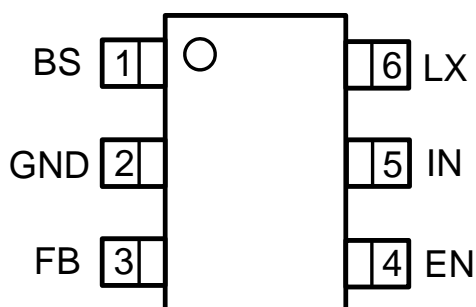
Pin Functions

Pin	Name	Function
1	BS	Bootstrap. A capacitor connected between LX and BS pins is required to form a floating supply across the high-side switch driver.
2	GND	Ground Pin
3	FB	Output Voltage feedback input. Connect FB to the center point of the external resistor divider.
4	EN	Drive this pin to a logic-high to enable the IC. Drive to a logic-low to disable the IC and enter micro-power shutdown mode. Don't floating EN.
5	IN	Power supply Pin
6	LX	Switching Pin

ESD Rating

Items	Description	Value	Unit
V_{ESD_HBM}	Human Body Model for all pins	±2000	V
V_{ESD_CDM}	Charged Device Model for all pins	±2000	V

JEDEC specification JS-001

Recommended Operating Conditions

Items	Description	Min	Max	Unit
Voltage Range	IN	4.5	30	V
T_J	Operating Junction Temperature	-40	125	°C

Thermal Resistance (Note3)

Items	Description	Value	Unit
θ_{JA}	Junction-to-ambient thermal resistance	120	°C/W
θ_{JC}	Junction-to-case(top) thermal resistance	45	°C/W

Electrical Characteristics

V_{IN}=12V, V_{OUT}=3.3V, T_A = 25°C, unless otherwise noted.

Parameter	Conditions	Min	Typ	Max	Unit
Input Voltage Range		4.5		30	V
OVP Threshold			31.5		V
UVLO Rising Threshold			4.3	4.5	V
UVLO Hysteresis			0.3		V
Quiescent Current	V_{EN}=2V, V_{FB}=V_{REF} x 105%		350	500	µA
Standby Current	V_{EN}=2V, Switching, No Load		8	14	mA
Shutdown Current	V_{IN}=12V, EN=0V		5	10	µA
Regulated Feedback Voltage	T_A=25°C	0.784	0.8	0.816	V
FB Input Leakage Current		-0.5		0.5	µA
High-Side Switch On-Resistance			250		mΩ
Low-Side Switch On-Resistance			120		mΩ
High-Side Switch Leakage Current	V_{EN}=0V, V_{LX}=0V	1		10	µA
Switch Current Limit (Note 4)	Minimum Duty Cycle		2.0		A
Oscillation Frequency		1.2	1.5	1.8	MHz
Maximum Duty Cycle			78		%
Minimum On-Time			100		ns
Minimum Off-Time			125		ns
Soft Start Time		0.5	0.7	1.0	ms
Hiccup on Time			1		ms
Hiccup Time Before Restart			4.5		ms
EN Rising Threshold		0.85	1.1	1.3	V
EN Hysteresis			150		mV
EN Enable Delay Time (Note 4)			180		µs
EN Input Leakage Current	V_{EN}=2V	-1		1	µA
Thermal Shutdown Threshold (Note 4)			165		°C
Thermal Shutdown Hysteresis (Note 4)			15		°C

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: T_J is calculated from the ambient temperature T_A and power dissipation P_D according to the following formula: $T_J = T_A + P_D \times \theta_{JA}$. The maximum allowable continuous power dissipation at any ambient temperature is calculated by $P_{D\,(MAX)} = (T_{J(MAX)} - T_A)/\theta_{JA}$.

Note 3: Measured on JESD51-7, 4-layer PCB.
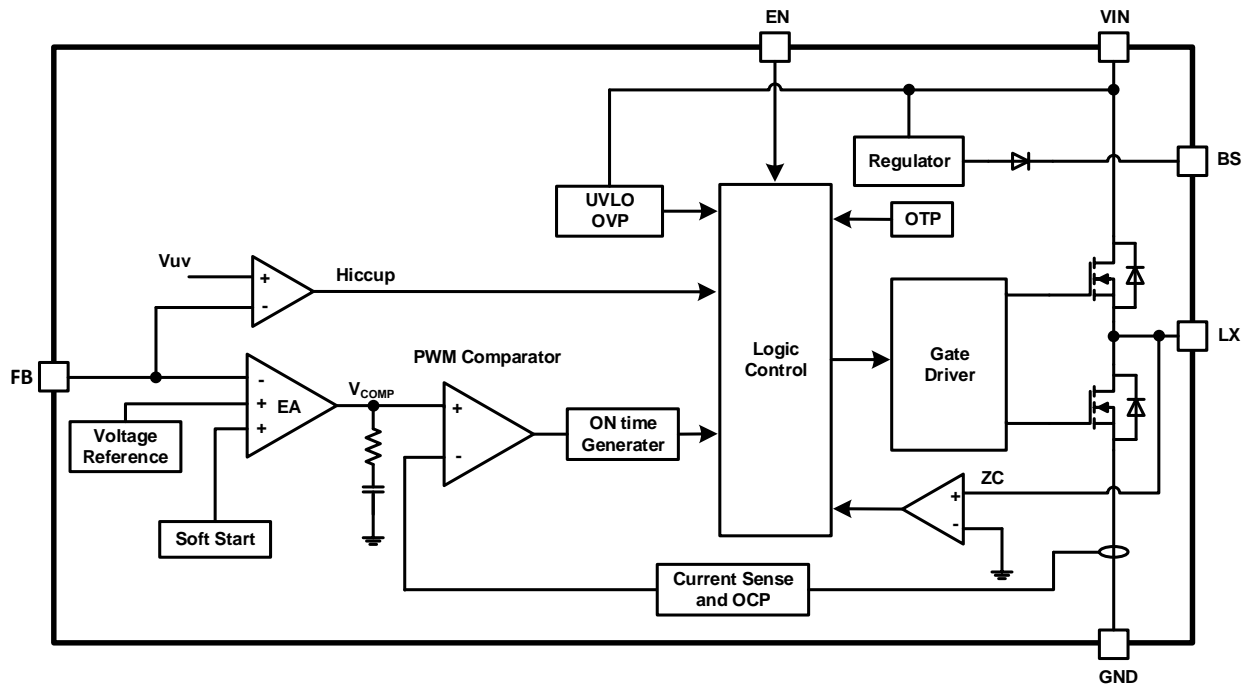
Note 4: Guaranteed by design.

Block Diagram

Figure 1. SL2359   **Block Diagram**

Operation Description

Internal Regulator

The SL2359 is a synchronous step down DC/DC converter with COT (constant on time) control that provides excellent transient response with no extra external compensation components. This device contains low resistance, high voltage high side power MOSFETs, and operates at 1.5MHz switching frequency and force PWM operating to ensure a compact, high efficiency design with excellent AC and DC performance.

Internal Soft-Start

The soft-start is implemented to prevent the converter output voltage from overshooting during startup. When the chip starts, the internal circuitry generates a soft-start voltage (SS) ramping up from 0V to V_{REF}. When it is lower than the internal FB reference (V_{REF}), SS overrides REF so the error amplifier uses SS as the reference. When SS is higher than V_{REF}, V_{REF} regains control. The SS time is internally fixed to 0.7ms typically. During the soft start process, SL2359 operates in PFM condition in order to achieve pre-bias start up function. After soft start, SL2359 switches operation condition to force PWM mode in all load range. The force PWM mode provides low output voltage ripples.

Over-Current-Protection and Short Circuits Protection

The SL2359 has cycle-by-cycle valley current limit function. When the inductor current value is larger than the current limit during MOSFET on state, the device enters into over current protection mode and MOSFET keeps on state until inductor current drops down to the value equal or lower than the current limit, and then high side MOSFET could turn on again.

If the output is short to GND and the output voltage drop until feedback voltage V_{FB} is below the output under-voltage V_{UV} threshold which is typically 50% of V_{REF}, SL2359 enters into hiccup mode to periodically disable and restart switching operation. The hiccup mode helps to reduce power dissipation and thermal rise during output short condition. The period of SL2359 hiccup mode is typically 5.5ms.

Startup and Shutdown

If both VIN and EN are higher than their appropriate thresholds, the chip starts switching operation. The reference block starts first, generating stable reference voltage and currents, and then the internal shut down the chip: EN low, VIN low and thermal shutdown. In the shutdown procedure, the signaling regulator is enabled. The regulator provides stable supply for the remaining circuitries. Three events can path is first blocked to avoid any fault triggering. The V_{COMP} voltage and the internal supply rail are then pulled down. Please do not float EN pin in the application.

Application Information

Setting the Output Voltage

The external resistor divider is used to set the output voltage (see Typical Application on page 1). Choose R1 to be around 39kΩ for optimal transient response. R2 is then given by:

Table 1: Selection for Common Output Voltages (V_{FB}=0.8V)

V_{OUT} (V)	R1 (kΩ)	R2 (kΩ)	C_{FB}(pF)	L (µH)
12	39	2.74	27	10
5	39	7.43	27	4.7
3.3	39	12.5	27	4.7
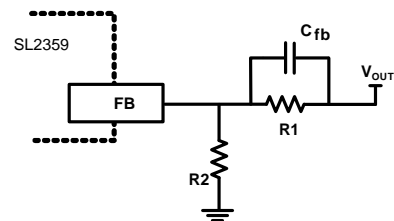
Figure 2. Feedback Network

SL2359 adopts COT control mode with 1.5MHz high switching frequency. The minimum off time of SL2359 is 100ns, so the minimum duty cycle with 1.5MHz switching frequency is 15%. When the operation duty cycle is lower than 15%, the switching frequency reduce. The recommended operation duty cycle is the range of 10% to 78%.

Selecting the Inductor

An inductor with a DC current rating of at least 25% percent higher than the maximum load current is recommended for most applications. For highest efficiency, the inductor DC resistance should be as small as possible. For most designs, the inductance value can be derived from the following equation.

$$L = \frac{V_{out} \times (V_{in} - V_{out})}{V_{in} \times \Delta I_L \times f_{OSC}}$$

Where ΔI_L is the inductor ripple current. Choose inductor ripple current to be approximately 30% if the maximum load current 1.2A. The maximum inductor peak current is:

$$I_{L(MAX)} = I_{LOAD} + \frac{\Delta I_L}{2}$$

Under light load conditions below 100mA, larger inductance is recommended for improved efficiency.

Selecting the Output Capacitor

The output capacitor (Co1) is required to maintain the DC output voltage. Ceramic, tantalum, or low ESR electrolytic capacitors are recommended. Low ESR capacitors are preferred to keep the output voltage ripple low. The output voltage ripple can be estimated by:
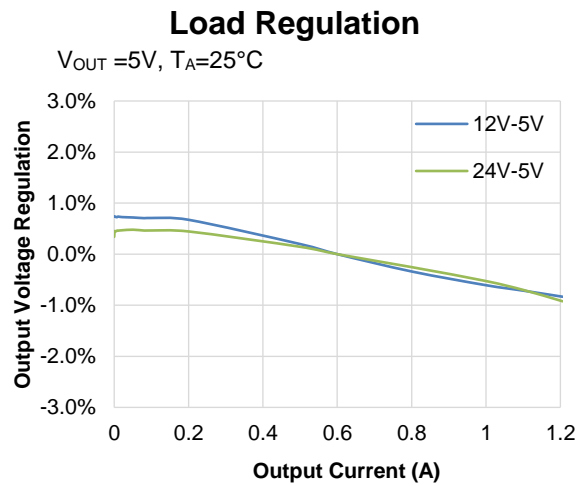
$$\Delta V_{OUT} = \frac{V_{OUT}}{f_S \times L} \times \left[1 - \frac{V_{OUT}}{V_{IN}}\right] \times \left[R_{ESR} + \frac{1}{8 \times f_S \times C_2}\right]$$

Where L is the inductor value and R_{ESR} is the equivalent series resistance (ESR) value of the output capacitor. In the case of ceramic capacitors, the impedance at the switching frequency is dominated by the capacitance. The output voltage ripple is mainly caused by the capacitance. For simplification, the output voltage ripple can be estimated by:
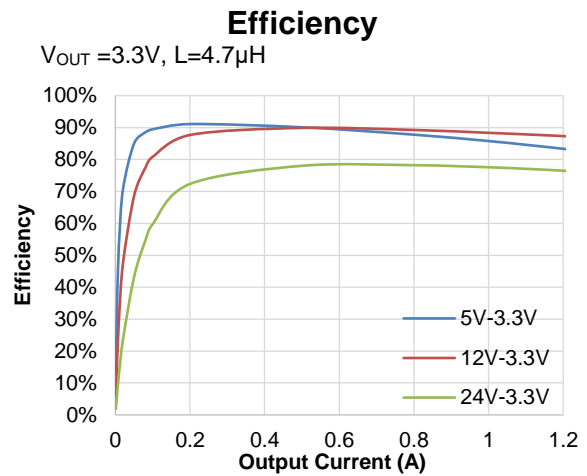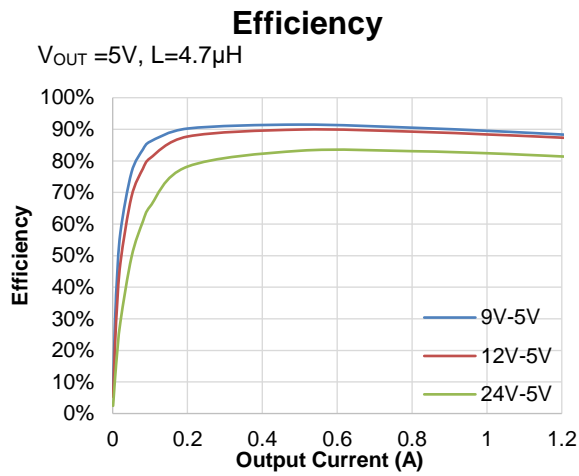
$$\Delta V_{OUT} = \frac{V_{OUT}}{8 \times f_S{}^2 \times L \times C_2} \times \left[1 - \frac{V_{OUT}}{V_{IN}}\right]$$

In the case of tantalum or electrolytic capacitors, the ESR dominates the impedance at the switching frequency. For simplification, the output ripple can be approximated to:

$$\Delta V_{OUT} = \frac{V_{OUT}}{f_S \times L} \times \left[1 - \frac{V_{OUT}}{V_{IN}}\right] \times R_{ESR}$$

The characteristics of the output capacitor also affect the stability of the regulation system. The SL2359 can be optimized for a wide range of capacitance and ESR values. 22µF or larger output capacitance is recommended for almost application. Because of DC derating of ceramic capacitor, the output capacitor DC voltage rating and package size must be considered in high output voltage application.
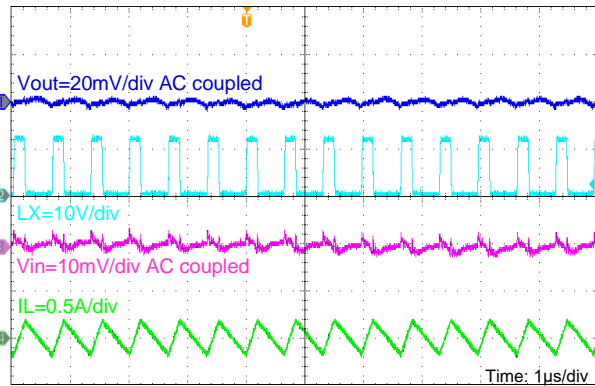
Typical Performance Characteristics

Efficiency

V_{OUT} =5V, L=4.7μH

Legend: 9V-5V, 12V-5V, 24V-5V

Efficiency

V_{OUT} =3.3V, L=4.7μH

Legend: 5V-3.3V, 12V-3.3V, 24V-3.3V

Efficiency

V_{OUT} =12V, L=10μH

Legend: 18V-12V, 24V-12V, 30V-12V

Load Regulation

V_{OUT} =5V, T_A=25°C

Legend: 12V-5V, 24V-5V

Load Regulation

V_{OUT} =3.3V, T_A=25°C

Legend: 12V-3.3V, 24V-3.3V

Vin vs. Switching Frequency

T_A=25°C, I_o= 0A

Legend: Vout=3.3V, Vout=5V
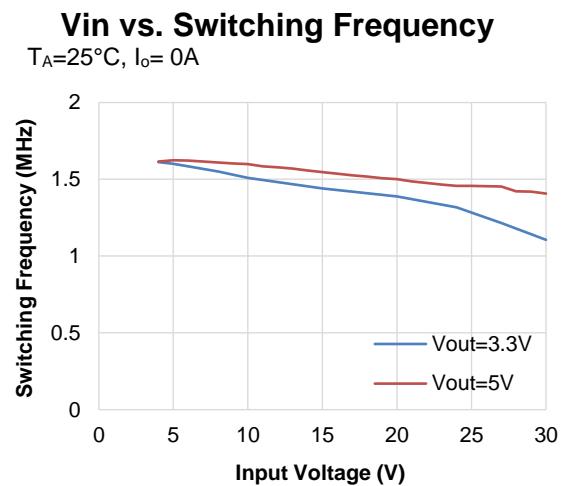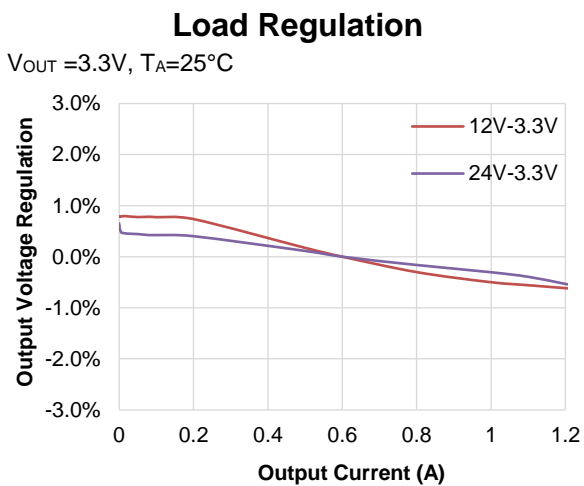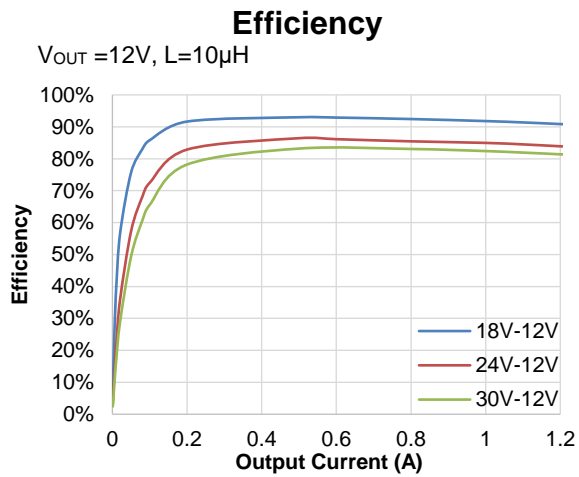
Typical Performance Characteristics(continued)

Steady State Operation
V_{IN} =12V, V_{OUT} =3.3V, No Load

Vout=20mV/div AC coupled
LX=10V/div
Vin=10mV/div AC coupled
IL=0.5A/div
Time: 1µs/div

Steady State Operation
V_{IN} =12V, V_{OUT} =3.3V, I_o= 1.2A

Vout=20mV/div AC coupled
LX=10V/div
Vin=50mV/div AC coupled
IL=1A/div
Time: 1µs/div
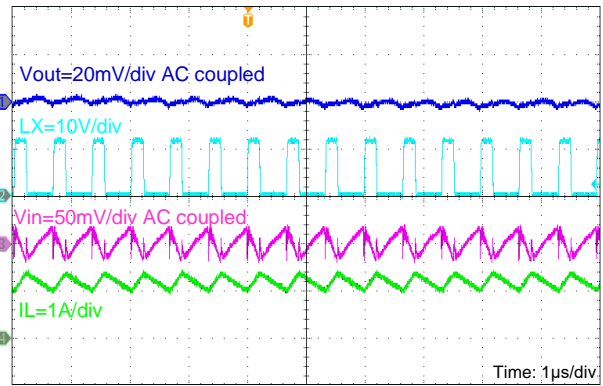
Load Transient
V_{IN} =12V, V_{OUT} =3.3V, I_o =0A to 0.6A

Vout=50mV/div AC coupled
Io=0.5A/div
Time: 200µs/div
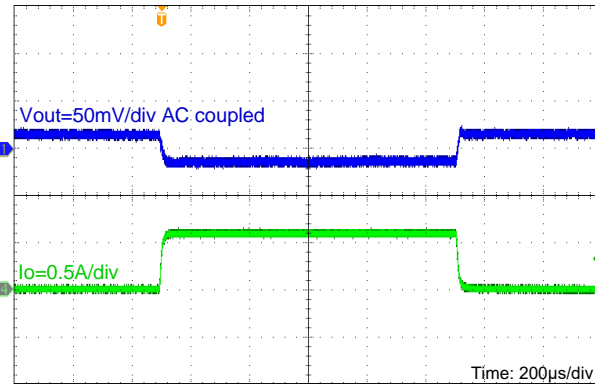
Load Transient
V_{IN} =12V, V_{OUT} =3.3V, I_o =0.6A to 1.2A
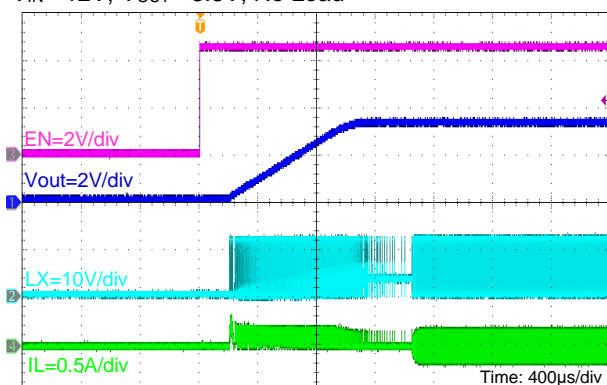

Vout=50mV/div AC coupled
Io=0.5A/div
Time: 200µs/div
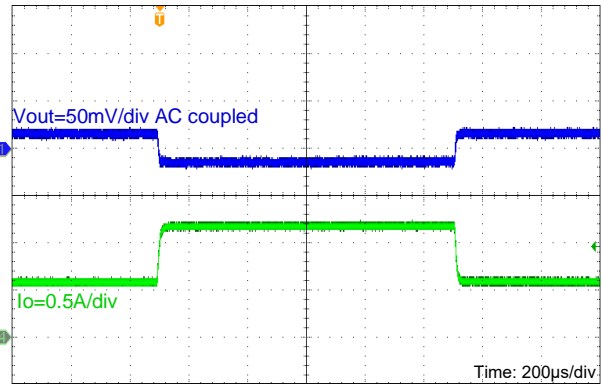
Output Short Entry
V_{IN} =12V, V_{OUT} =3.3V, I_o = 1.2A

Vin=10V/div
Vout=2V/div
LX=10V/div
IL=2A/div
Time: 4ms/div
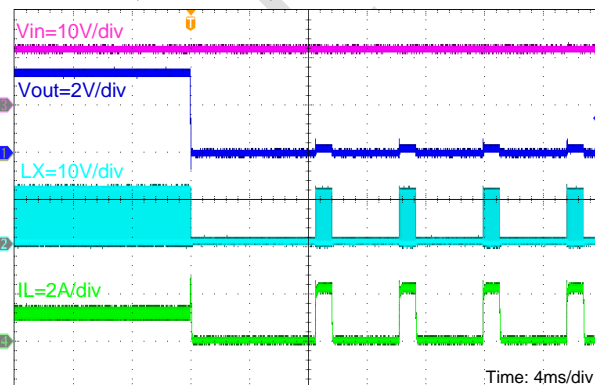
Output Short Recovery
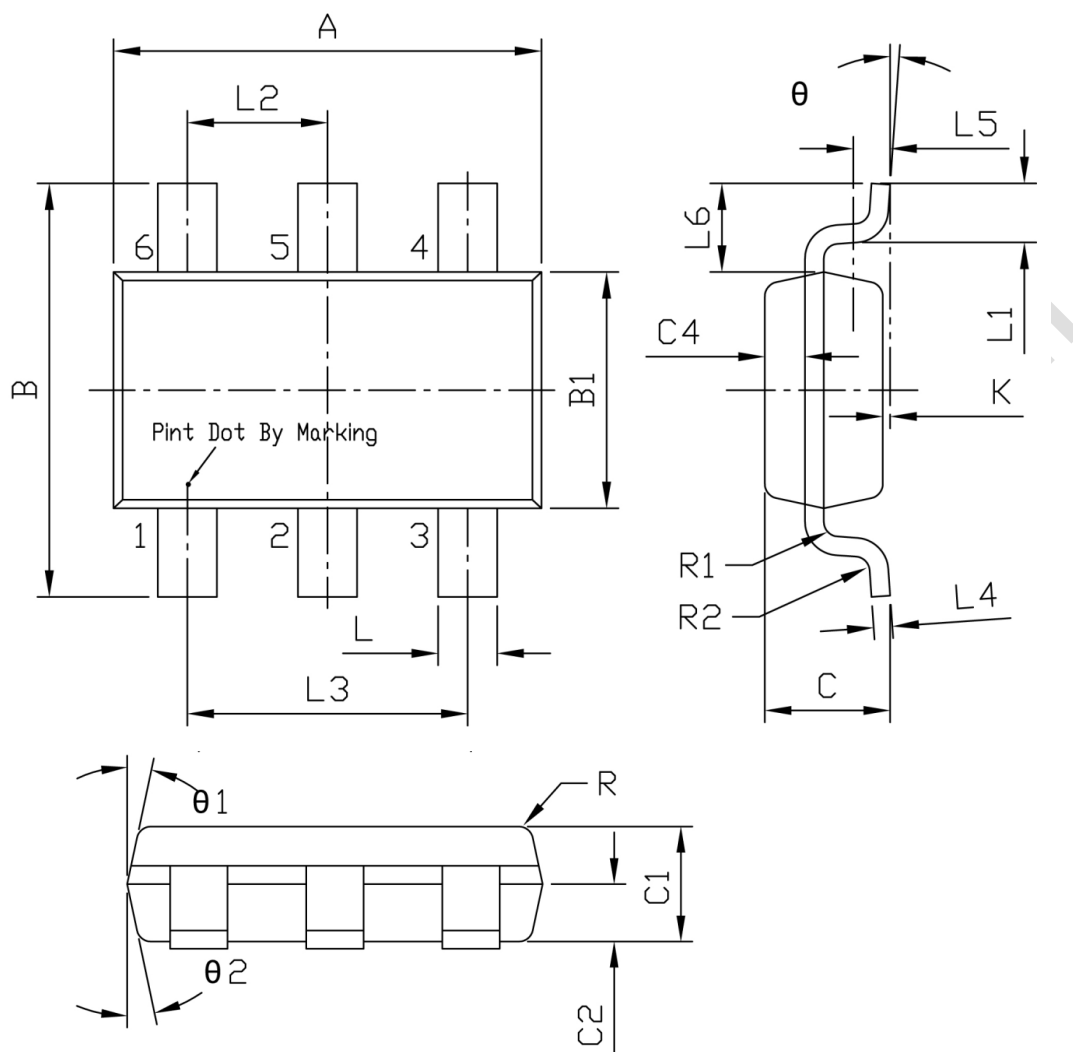V_{IN}=12V, V_{OUT} =3.3V, I_o = 1.2A

Time:4ms/div
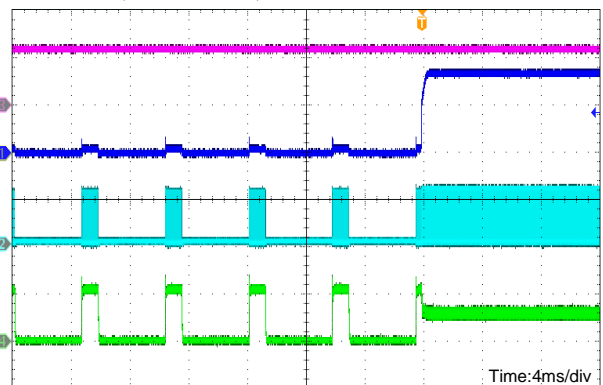
Typical Performance Characteristics(continued)

Input Power On
V_{IN} =12V, V_{OUT} =3.3V, I_o = 1.2A

Vin=5V/div
Vout=2V/div
LX=10V/div
IL=2A/div
Time: 400µs/div
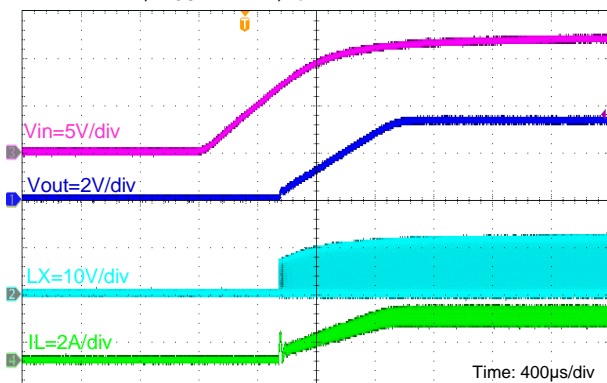
Input Power Down
V_{IN} =12V, V_{OUT} =3.3V, I_o = 1.2A
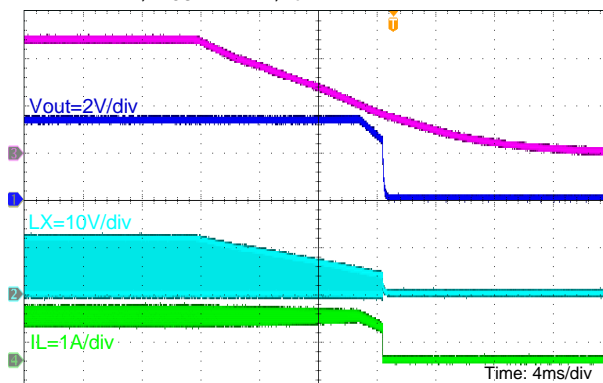
Vout=2V/div
LX=10V/div
IL=1A/div
Time: 4ms/div

EN Enable
V_{IN} =12V, V_{OUT} =3.3V, No Load

EN=2V/div
Vout=2V/div
LX=10V/div
IL=0.5A/div
Time: 400µs/div

EN Disable
V_{IN} =12V, V_{OUT} =3.3V, No Load EN=2V/div

Vout=2V/div
LX=10V/div
IL=1A/div
Time: 400ms/div

EN Enable
V_{IN} =12V, V_{OUT} =3.3V, I_o = 1.2A

EN=2V/div
Vout=2V/div
LX=10V/div
IL=1A/div
Time: 400µs/div

EN Disable
V_{IN} =12V, V_{OUT} =3.3V, I_o = 1.2A

Vout=2V/div
LX=10V/div
IL=1A/div
Time: 100µs/div
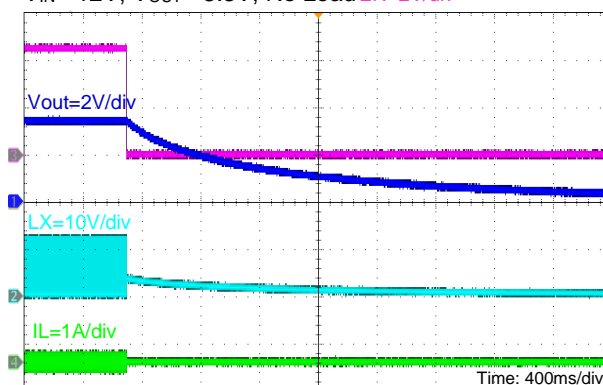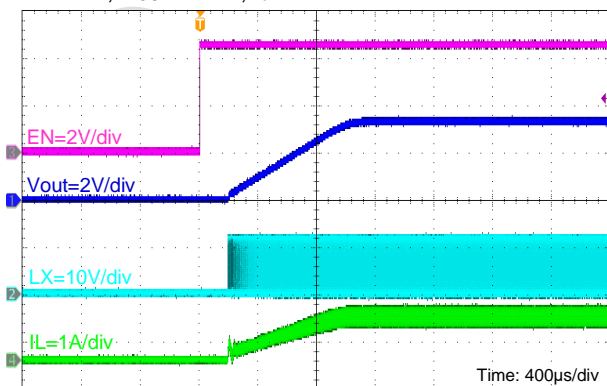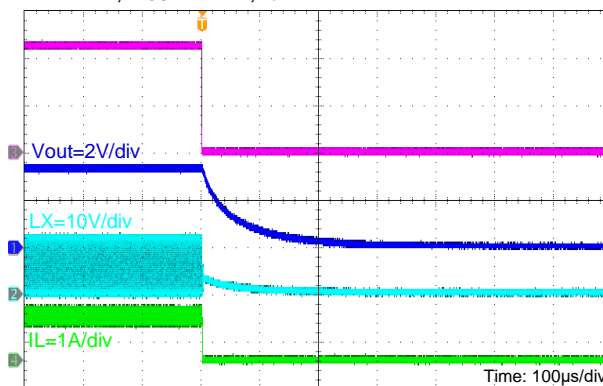
Package Information

SOT23-6

Unit: mm

Symbol	Dimensions In Millimeters			Symbol	Dimensions In Millimeters		
	Min	Typ	Max		Min	Typ	Max
A	2.80	2.90	3.00	L3	1.800	1.900	2.000
B	2.60	2.80	3.00	L4	0.077	0.127	0.177
B1	1.50	1.60	1.70	L5	-	0.250	-
C	-	-	1.05	L6	-	0.600	-
C1	0.60	0.80	1.00	θ	0°		0°
C2	0.35	0.40	0.45	θ1	10°	12°	14°
C4	0.223	0.273	0.323	θ2	10°	12°	14°
K	0.000	0.075	0.150	R	-	0.100	-
L	0.325	0.400	0.475	R1	-	0.100	-
L1	0.325	0.450	0.550	R2	-	0.100	-
L2	0.850	0.950	1.050				